Time-Division Transmission is Optimal for Covert Communication over Some Broadcast Channels

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Consider a traditional two-user broadcast channel

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■ Should not be able to distinguish between Q_{Zⁿ} (output dist. induced by a code) and the "no-communication" output dist. Q₀^{×n}.

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- Warden attempts to design optimal detector to distinguish
 - H₀: observed distribution is $Q_0^{\times n}$ (no communication)
 - H₁: observed distribution is \hat{Q}_{Z^n} (communication active)

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- Optimal performance

$$\pi_{1|0} + \pi_{0|1} = 1 - \frac{1}{2} \left\| Q_0^{\times n} - \hat{Q}_{Z^n} \right\|_1 \ge 1 - \sqrt{D(\hat{Q}_{Z^n} \| Q_0^{\times n})}$$

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For covert communication, we want to make $D(\hat{Q}_{Z^n} || Q_0^{\times n})$ small.

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Context

Growing concern for privacy and confidentiality

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Context

- Growing concern for privacy and confidentiality
- Renewed interest in fundamental limits of covert communications:
 - Secure space-time codes [Hero '03]
 - Secure stegosystems [Korzhik et al. '05]
 - $O(\sqrt{n})$ bits over *n* ch. uses with $O(\sqrt{n}\log n)$ key [Bash et al. '12]
 - Similar to square-root law in steganography [Cachin '04]

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 - $O(\sqrt{n})$ bits over *n* ch. uses with $O(\sqrt{n} \log n)$ key [Bash et al. '12]
 - Similar to square-root law in steganography [Cachin '04]
- Several extensions and related results:
 - Constants in $O(\sqrt{n})$ term [Wang, Wornell, Zheng '16 and Bloch '16]
 - Second-order [Tahmasbi-Bloch '16]
 - Error exponents [Tahmasbi-Bloch-Tan '17]
 - Multi-user [Arumugam-Bloch '16, '17]

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Definition of a code

An $(n, M_{1n}, M_{2n}, \varepsilon, \delta)$ -code for the broadcast channel with a warden $P_{Y_1, Y_2, Z|X}$ consists of

- Two message sets $\mathcal{M}_j := \{1, \ldots, M_{jn}\}$ for j = 1, 2;
- Two independent messages uniformly distributed over their respective message sets, i.e., $W_j \sim \text{Unif}(\mathcal{M}_j)$ for j = 1, 2;

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• One encoder
$$f: \mathcal{M}_1 \times \mathcal{M}_2 \to \mathcal{X}^n$$
;

• Two decoders $\varphi_j : \mathcal{Y}_j^n \to \mathcal{M}_j$ for j = 1, 2;

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such that the following constraints hold:

$$\mathsf{Reliability:} \qquad \Pr\left(\cup_{j=1}^2 \{\hat{W}_j \neq W_j\}\right) \leq \varepsilon$$

and

Covertness:
$$D(\hat{Q}_{Z^n} || Q_0^{\times n}) \leq \delta.$$

Definition of Covert Capacity Region

■ $(L_1, L_2) \in \mathbb{R}^2_+$ is (ε, δ) -achievable if there exists a sequence of $(n, M_{1n}, M_{2n}, \varepsilon_n, \delta)$ -codes such that

$$\liminf_{n \to \infty} \frac{1}{\sqrt{n\delta}} \log M_{jn} \ge L_j, \quad j = 1, 2,$$
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The δ -covert capacity region

$$\mathcal{L}_{\delta} := igcap_{arepsilon \in (0,1)} \mathcal{L}_{arepsilon,\delta} = \lim_{arepsilon o 0} \mathcal{L}_{arepsilon,\delta}.$$

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For simplicity, assume binary-input channels, i.e., $\mathcal{X} = \{0, 1\}$

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Definition of Covert Capacity

- For simplicity, assume binary-input channels, i.e., $\mathcal{X} = \{0, 1\}$
- Given a DMC with a warden $P_{Y,Z|X}$, the covert capacity [Wang, Wornell, Zheng '16 and Bloch '16] is

$$L^{*}(P_{Y,Z|X}) := \sqrt{\frac{2D(W(\cdot|1)||W(\cdot|0))^{2}}{\chi_{2}(Q_{1}||Q_{0})}}$$

where

$$W(\cdot|x) = P_{Y|X}(\cdot|x) \qquad Q_x := P_{Z|X}(\cdot|x), \quad \forall x \in \mathcal{X},$$

and

$$\chi_2(\mathcal{Q}_1 \| \mathcal{Q}_0) := \sum_z \frac{(\mathcal{Q}_1(z) - \mathcal{Q}_0(z))^2}{\mathcal{Q}_0(z)}.$$

Assumption

Condition 1: Fix a BC with a warden $P_{Y_1,Y_2,Z|X}$.

Let the covert capacities of $P_{Y_1,Z|X}$ and $P_{Y_2,Z|X}$ be L_1^* and L_2^* respectively.

If $L_1^* \ge L_2^*$ assume that

$$\max_{P_X} \frac{I(X;Y_1)}{I(X;Y_2)} \le \frac{L_1^*}{L_2^*}.$$

Otherwise, if $L_1^* \leq L_2^*$ assume that

$$\max_{P_X} \frac{I(X;Y_2)}{I(X;Y_1)} \le \frac{L_2^*}{L_1^*}.$$

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Discussion of Condition 1

Easy to check for binary-input BCs:

$$\frac{D(W_1 \| W_0)}{D(V_1 \| V_0)} \le \min_{\gamma \in [0,1]} \frac{D(W_\gamma \| W_0)}{D(V_\gamma \| V_0)}, \qquad W_\gamma(y) = \sum_{x \in \mathcal{X}} P_\gamma(x) W(y|x).$$

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Let $W = P_{Y_1|X} = BSC(p)$ and $V = P_{Y_2|X} = \begin{bmatrix} 1 - q_0 & q_0 \\ q_1 & 1 - q_1 \end{bmatrix}$



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Covert Broadcast Communication

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Main Result

Theorem (Tan-Lee (2018))

Assume Condition 1 holds for $P_{Y_1,Y_2,Z|X}$. For any $\delta > 0$ and $P_{Y_j|X}(\cdot|1) \ll P_{Y_j|X}(\cdot|0)$ for j = 1, 2,

$$\mathcal{L}_{\delta} = \left\{ (L_1, L_2) \in \mathbb{R}^2_+ : \frac{L_1}{L_1^*} + \frac{L_2}{L_2^*} \le 1
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Effect of Key Size

Theorem (Tan-Lee (2018))

Under Condition 1, the tuple (L_1, L_2, L_{key}) is achievable if and only if

$$\frac{L_1}{L_1^*} + \frac{L_2}{L_2^*} \le 1$$

and

$$L_{\text{key}} \ge \left(rac{L_1}{L_1^*} + rac{L_2}{L_2^*}
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where

$$L_Z^* = L^*(P_{Z,Z|X})$$

is the self-covert capacity of the channel $X \rightarrow Z$.

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If we operate on the boundary of the keyless covert capacity region,

$$L_{\rm key}^* = L_Z^* - L_1 - L_2$$

Time-Division is Optimal over Some BCs



■ Use an optimal code for $X \to Y_1$ for ρn channel uses. If $\delta' < \delta$,

$$\log M_{1n} \cong \sqrt{\rho n \delta'} L_1^*$$

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■ Use another optimal code for $X \rightarrow Y_2$ over $(1 - \rho)n$ uses. Then,

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• Choose $\delta' = (1 - \rho)\delta$ achieves the point $(\rho L_1^*, (1 - \rho)L_2^*)$.

Time-Division is Optimal for Some BCs: Why?

■ Covert communication implies that Xⁿ must have low weight of order Θ(¹/_{√n}) [Wang, Wornell, Zheng '16 and Bloch '16], i.e.,

$$|\{i \in [n] : X_i = 1\}| = \Theta(\sqrt{n})$$

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- For illustration purposes, consider a BS-BC $P_{Y_1,Y_2|X}$ is such that $P_{Y_i|X}$ for j = 1, 2 are BSCs.
- Same intuition for Gaussian broadcast channels
- But use Entropy Power Inequality instead of Mrs. Gerber's Lemma.

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Superposition coding: Cloud center $u_2^n(w_2)$ carries message w_2 ; Satellite codeword $x^n(w_1, w_2) = u_1^n(w_1) \oplus u_2^n(w_2)$ carries (w_1, w_2)

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Since $x^n(w_1, w_2)$ has low weight (say α_n) and $u_1^n(w_1)$ and $u_2^n(w_2)$ are randomly chosen, locations of 1's in $u_1^n(w_1)$ and $u_2^n(w_2)$ are not likely to overlap.

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Assume weight of $u_1^n(w_1)$ is $\rho\alpha_n$ and that of $u_2^n(w_2)$ is $(1-\rho)\alpha_n$

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Consider BSBCs

$$Y_1 = X \oplus N_1, \qquad Y_2 = X \oplus N_2, \quad N_j \sim \operatorname{Bern}(p_j), \quad p_2 \ge p_1$$

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Put

wt $(u_1^n(w_1)) = \rho \alpha_n n$, and wt $(u_2^n(w_2)) = (1 - \rho) \alpha_n n$, the superposition coding inner bound with $X = U_1 \oplus U_2$ reads $(U_2 - X - Y_1 - Y_2)$

$$R_1 \le I(X; Y_1 | U_2) = I(U_1; Y_1 \oplus N_1) \approx \rho \alpha_n L_1^*$$

$$R_2 \le I(U_2; Y_2) = I(U_2; U_2 \oplus \tilde{N}_2) \approx (1 - \rho) \alpha_n L_2^*$$

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$$R_1 \leq I(X; Y_1 | U_2) = I(U_1; Y_1 \oplus N_1) \approx \rho \alpha_n L_1^*$$

$$R_2 \leq I(U_2; Y_2) = I(U_2; U_2 \oplus \tilde{N}_2) \approx (1 - \rho) \alpha_n L_2^*$$

Hence, we can write

$$\frac{R_1}{L_1^*} + \frac{R_2}{L_2^*} \lessapprox \alpha_n$$

El Gamal's Converse for More Capable BCs

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El Gamal's Converse for More Capable BCs

Lemma (El Gamal (1979))

Every $(n, M_{1n}, M_{2n}, \varepsilon_n)$ -code for any BC satisfies

$$(\log M_{1n})(1 - \varepsilon_n) - 1 \le \sum_{i=1}^n I(U_{1i}; Y_{1i})$$

 $(\log M_{2n})(1 - \varepsilon_n) - 1 \le \sum_{i=1}^n I(U_{2i}; Y_{2i})$

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$$(\log M_{1n} + \log M_{2n})(1 - \varepsilon_n) - 2 \le \sum_{i=1}^n \left[I(X_i; Y_{1i} | U_{2i}) + I(U_{2i}; Y_{2i}) \right]$$
$$(\log M_{1n} + \log M_{2n})(1 - \varepsilon_n) - 2 \le \sum_{i=1}^n \left[I(U_{1i}; Y_{1i}) + I(X_i; Y_{2i} | U_{1i}) \right]$$

where U_{1i} and U_{2i} satisfy $(U_{1i}, U_{2i}) - X_i - (Y_{1i}, Y_{2i})$.

Assume $L_1^* \ge L_2^*$ (wlog) and let

$$\lambda = \frac{L_1^*}{L_2^*} \ge 1.$$

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Assume $L_1^* \ge L_2^*$ (wlog) and let

$$\lambda = \frac{L_1^*}{L_2^*} \ge 1.$$

 Combining previous inequalities and using a standard time-sharing random variable, we obtain

$$\frac{1}{n} \left[\log M_{1n} + \lambda \log M_{2n} - (1+\lambda) \right] \le \max_{U,X} I(X;Y_1|U) + \lambda I(U;Y_2)$$

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- Problem: Maximization of $I(X; Y_1|U) + \lambda I(U; Y_2)$ over all (U, X) requires tools specific to the broadcast channel
- For the BS-BC, Mrs. Gerber's Lemma [Wyner-Ziv (1973)] helps to simplify

Remove U's by exploiting tools from convex analysis

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Remove U's by exploiting tools from convex analysis

■ Note that $U - X - Y_2$ forms a Markov chain so

$$\begin{split} & \max_{P_{U,X}} I(X;Y_1|U) + \lambda I(U;Y_2) \\ & = \max_{P_{U,X}} I(X;Y_1|U) + \lambda [I(X;Y_2) - I(X;Y_2|U)] \\ & = \max_{P_X} \lambda I(X;Y_2) + \max_{P_{U|X}} [I(X;Y_1|U) - \lambda I(X;Y_2|U)] \end{split}$$

Remove U's by exploiting tools from convex analysis

■ Note that $U - X - Y_2$ forms a Markov chain so

$$\max_{P_{U,X}} I(X; Y_1|U) + \lambda I(U; Y_2)$$

= $\max_{P_{U,X}} I(X; Y_1|U) + \lambda [I(X; Y_2) - I(X; Y_2|U)]$
= $\max_{P_X} \lambda I(X; Y_2) + \max_{P_{U|X}} [I(X; Y_1|U) - \lambda I(X; Y_2|U)]$

Now,

$$\max_{P_{U|X}} [I(X;Y_1|U) - \lambda I(X;Y_2|U)] = \mathbb{C}[I(P_X,W) - \lambda I(P_X,V)]$$

where $W = P_{Y_1|X}$ and $V = P_{Y_2|X}$ and the concave envelope is defined as

$$\mathbb{C}[f](x) := \inf\{g(x) : g \ge f, g \text{ is concave}\}\$$

The usual superposition coding region is

$$\mathcal{C} = \bigcup_{P_X, P_{U|X}} \left\{ (R_1, R_2) \in \mathbb{R}^2_+ \mid R_1 \le I(X; Y_1|U), R_2 \le I(U; Y_2) \right\}$$

Using the concave envelope representation, we have

$$\mathcal{C} = \bigcap_{\lambda \ge 1} \left\{ (R_1, R_2) \in \mathbb{R}^2_+ \mid R_1 + \lambda R_2 \le \max_{P_X} \lambda I(X; Y_2) + T_\lambda(X) \right\}$$

where $T_{\lambda}(X) := \mathbb{C}[I(X; Y_1) - \lambda I(X; Y_2)].$

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Covert Broadcast Communication

The usual superposition coding region is

$$\mathcal{C} = \bigcup_{P_X, P_{U|X}} \left\{ (R_1, R_2) \in \mathbb{R}^2_+ \mid R_1 \le I(X; Y_1|U), R_2 \le I(U; Y_2) \right\}$$

Using the concave envelope representation, we have

$$\mathcal{C} = \bigcap_{\lambda \ge 1} \left\{ (R_1, R_2) \in \mathbb{R}^2_+ \mid R_1 + \lambda R_2 \le \max_{P_X} \lambda I(X; Y_2) + T_\lambda(X) \right\}$$

where $T_{\lambda}(X) := \mathbb{C}[I(X; Y_1) - \lambda I(X; Y_2)].$



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$$\frac{\log M_{1n} + \lambda \log M_{2n}}{n} \lesssim \max_{P_X} \lambda \cdot I(P_X, V) + \mathbb{C} \big[I(P_X, W) - \lambda \cdot I(P_X, V) \big]$$

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- Finally, recalling that $\lambda = L_1^*/L_2^*$,

$$L_1 + \frac{L_1^*}{L_2^*} \cdot L_2 \lessapprox \frac{L_1^*}{L_2^*} \cdot L_2^* = L_1^*, \quad \Longrightarrow \quad \frac{L_1}{L_1^*} + \frac{L_2}{L_2^*} \le 1.$$

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Conclusion and Open Problems



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- What can we say about BCs which don't satisfy Condition 1?
- More than 2 legitimate receivers?

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The Department of Electrical and Computer Engineering (ECE) at the National University of Singapore is offering positions for postdoctoral fellows who will work in information theory, machine learning and their intersection.

The Department of Electrical and Computer Engineering (ECE) at the National University of Singapore (NUS) is offering positions for postdoctoral fellows who will work closely with Dr. Vincent Tan at the intersection of information theory, statistical signal processing, and machine learning. Some sample topics include:

- Fundamental performance limits (and algorithms) for dictionary learning (e.g., matrix factorization), ranking, and deep learning architectures;
- Learning in the presence of privacy constraints;
- Learning in the large alphabet regime;
- · Learning of graphical models and other statistical models.

Working in traditional topics in Shannon's information theory of interest to the PI will also be highly encouraged. Some sample topics include:

- Multi-user information theory;
- Strong converse and second-order asymptotics;
- Error exponent analysis and the method of types;
- Information-theoretic security;

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